

Description

Oversampling in analog/digital and digital/analog converters

5 Related Applications

This application is a continuation of PCT patent application number PCT/EP01/03685, filed March 30, 2001, the disclosure of which is incorporated herein by reference in its entirety.

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Technical Field

The present invention relates to analog/digital and digital/analog converters and, in particular, relates to a method for oversampling in analog/digital and  
15 digital/analog converters which are used for data transmission by means of digital subscriber lines (DSL) and very high data rate digital subscriber lines (VDSL).

Background Art

20 VDSL is the name for a transmission method, the transmission rate of which is higher than that of ADSL (Asymmetric Digital Subscriber Line). Conventionally, a VDSL system is used for a hybrid network which consists of optical fiber and copper lines, the use of optical  
25 fiber cables, in particular, providing high transmission rates.

In a hybrid network, an optical fiber cable is run up to the switching offices in the local area or even up to  
30 the cable distribution boxes on the side of the road, such conventional applications being known, for example,

from the Internet address "http://www.e-online.de/sites/kom/0305237.htm.

An essential feature of digital transmission systems is  
5 an analog/digital conversion, digital transmission and  
subsequent digital/analog conversion of relevant data  
streams. In this arrangement, various methods of  
oversampling are conventionally used as described, for  
example, in the Internet reference "http://www.hoer-  
10 wege.de/over+upsamp.htm.

To increase an effective bit number in conventional  
methods for digital data transmission, a method of noise  
shaping is conventionally used.

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Conventional noise shaping which is used both for an A/D  
signal path and for a D/A signal path is described in  
the printed document "NORSWORTHY S R ET AL: "A  
programmable codec signal processor" SOLID-STATE  
20 CIRCUITS CONFERENCE, 1996. DIGEST OF TECHNICAL PAPERS.  
42ND ISSCC., 1996 IEEE INTERNATIONAL SAN FRANCISCO, CA,  
USA 8-10 FEB. 1996, NEW YORK, NY, USA, IEEE, US, 8th  
February 1996 (1996-02-08), pages 170-171, 438,  
XP010156441 ISBN: 0-7803-3136-2".

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This printed document shows that second-order noise  
shaping can be performed but, disadvantageously, no  
separate noise shaping devices are specified which could  
provide adapted noise shaping. In the conventional  
30 method described, it is not possible, disadvantageously,  
to selectively provide an allocation of noises in  
individual frequency bands.

Furthermore, it is not suitable that the first noise shaping provided in a first noise shaping device cannot be combined with a second noise shaping provided in a second noise shaping device.

Figure 3 shows a circuit arrangement of a conventional analog/digital converter as used, for example, in digital transmission systems. An analog input signal 100 is supplied to an input terminal 102. The analog/digital converter 101 contains a summing device 104 and a quantizing device 111 and a digital signal 105 can be picked up as output signal at an output terminal 106, and a noise signal 113 can be connected to a second input terminal, a noise source terminal 103.

The analog input signal 100 and a noise signal 113 connected to the noise source terminal 103 are superimposed in the summing device 104, the summed signal being supplied to the quantizing device 111. The output of the quantizing device 111 is connected to the output terminal 106, a digital signal 105 being provided as the output signal.

Such analog/digital converters according to the prior art have a number of disadvantages. In VDSL systems, analog/digital converters and digital/analog converters with an effective resolution of 9 bits to 12 bits are used. The resolution of these converters influences costs and chip area and energy consumption to a high degree in a design of an integrated circuit. On the other hand, converters with a high resolution allow the

complexity of external components such as duplex filters, which can be connected to the integrated circuit, to be reduced.

- 5 Thus, a balance must be struck between the costs of an integrated circuit (or of a chip area) and a complexity (thus also the costs) of external components in designing a circuit.

10 Summary of the Invention

It is thus an object of the present invention to provide analog/digital converters and digital/analog converters with high resolution without increasing an energy consumption and a chip area.

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This object is achieved by a method for the digital transmission of analog signals in which oversampling is performed, according to claim 1, and a device having the features of claim 12.

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The method according to the invention for the digital transmission of analog signals in which oversampling is performed, according to claim 1, and the device having the features of claim 12, respectively, have the

25 following advantages.

The method according to the invention is advantageously used together with a noise shaping concept in VDSL (Very High Data Rate DSL) systems which increases an effective  
30 bit number.

It is also advantageous that a chip area of a factor of about 2 is saved by increasing a resolution by 1 bit.

The core of the invention is a method for oversampling  
5 in analog/digital and digital/analog converters in which  
oversampling with a first noise shaping which is  
performed by means of a first noise shaping device  
during an analog/digital conversion, is combined with a  
second noise shaping which is performed by means of a  
10 second noise shaping device during a digital/analog  
conversion.

The subclaims contain advantageous developments and  
improvements of the respective subject matter of the  
15 invention.

According to yet another preferred development of the  
present invention, a second-order comb filter is used as  
the decimation filter unit.

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According to yet another preferred development of the  
present invention, the first noise shaping device is a  
first-order noise shaping device.

25 According to yet another preferred development of the  
present invention, the second noise shaping device is a  
first-order noise shaping device.

According to yet another preferred development of the  
30 present invention, an interpolation filter unit is a  
second-order comb filter.

According to yet another preferred development of the present invention, a digital/analog converter is a 10-bit current drive converter.

- 5 According to yet another preferred development of the present invention, noise shaping is provided by an adaptive noise shaping device.

Brief Description of the Drawings

- 10 Illustrative embodiments of the invention are shown in the drawings and explained in greater detail in the description following. In the drawings:

Figure 1a shows a circuit arrangement for  
15 analog/digital conversion according to an illustrative embodiment of the present invention;

Figure 1b shows a further circuit arrangement for  
20 analog/digital conversion according to a further illustrative embodiment of the present invention which, in comparison with the circuit arrangement shown in figure 1a, is extended by a first noise shaping device;

25 Figure 2 shows a circuit arrangement for digital/analog conversion with a second noise shaping device according to an illustrative embodiment of the present invention; and

30 Figure 3 shows a conventional circuit arrangement for analog/digital conversion.

Detailed Description of the Invention

Figure 1a shows a circuit arrangement for analog/digital conversion according to an illustrative embodiment of  
5 the present invention.

In the circuit arrangement shown in figure 1a, a conventional analog/digital converter 101 is used which is extended by a decimation filter unit 107. An output  
10 signal of the analog/digital converter 101 is provided as a digital signal 105 at an output terminal 106 and the decimation filter unit 107 is supplied with a digital signal 105 corresponding to an analog input signal 100.

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The analog input signal 100 is converted into the digital signal 105 in a conventional manner in that the analog input signal 100 is supplied to an input terminal 102 of the analog/digital converter 101. Furthermore, a  
20 noise signal 113 is supplied to a noise source terminal 103, the analog input signal 100 and the noise signal 113 supplied to the noise source terminal 103 being superimposed in a summing device 104. The output signal of the summing device 104 is quantized in a quantizing  
25 device 111 in order to provide the output signal as a digital signal 105.

According to an illustrative embodiment of the present invention, a decimation filter unit 107, in which low-  
30 pass filtering is performed, is connected to the output. The decimation filter unit 107 is provided, for example, by a digital low-pass filter which reduces a frequency

bandwidth of, for example, 260 MHz to a frequency bandwidth of 40 MHz, achieving an oversampling factor of approx. 6.6.

5 As shown in this illustrative embodiment, a resolution of the analog/digital converter is advantageously increased from 10 bits to 11 bits. As the output signal of the overall circuit arrangement shown in figure 1a, a digital transmission signal 110 is provided at the  
10 output of the decimation filter unit 107.

The circuit arrangement shown in figure 1b corresponds to the circuit arrangement shown in figure 1a, with the exception that a first noise shaping device 112 is  
15 arranged between the analog/digital converter 101 and the decimation filter unit 107.

Identical reference symbols as in figure 1a correspond to identical or similar components which are not  
20 explained again in order to avoid an overlapping description.

The first noise shaping device 112 is used for further increasing the resolution of the overall circuit  
25 arrangement. In the case of uniform quantization, identical noise energy (white noise) is in each case added to frequency sub-bands in the entire spectrum so that, during quantization, the least permissible distortion in the respective frequency band must always  
30 be used as a measure of the noise energy. In this arrangement, a maximum bit rate can disadvantageously not be utilized completely because specific frequency



sub-bands permit less noise energy than the remaining frequency sub-bands.

Noise shaping performed in the first noise shaping device 112 enables an allocation of noise in individual frequency bands to be specified.

Combining the first noise shaping device 112 with the oversampling provided by the decimation filter unit 107 offers considerable advantages, particularly for VDSL systems providing, for example, a reduction in chip area needed, a reduction in energy consumption, improved resolution etc.

Figure 2 illustrates a circuit arrangement for digital/analog conversion with a second noise shaping device 205 according to an illustrative embodiment of the present invention.

In the circuit arrangement shown in figure 2, digital/analog conversion is performed which, in particular, is suitable for VDSL systems. A transmitted digital transmission signal 110 is supplied to a mixing unit 201. In the mixing unit 201, a receive noise signal 211 applied to a receive noise source terminal 209 is superimposed on the digital transmission signal 110. The output signal of the mixing unit 201 is supplied to a post-quantizing device 202, an output signal quantized by the post-quantizing device 202 being supplied to an interpolation filter unit 203.

In the opposite way to the decimation filter unit 107 described with reference to figure 1a, b, an interpolation filter unit 203 provides an increase in the frequency bandwidth by the same factor by which the frequency bandwidth was reduced in the example shown in figure 1a, b, in this case by a factor which is approx. 6.6.

The output signal of the post-quantizing unit 202 is supplied to the interpolation filter unit 203 which provides suitable oversampling. An output signal of the interpolation filter unit 203 is supplied to an amplifier unit 204 in which the output signal of the interpolation filter unit 203 is amplified by a specifiable factor which, in particular, is suitable for driving the subsequent second noise shaping device 205.

An output signal of the amplifier unit 204 is supplied to the input of the second noise shaping device 205. In the second noise shaping device 205, noise shaping like that described with reference to the first noise shaping device 112 as in figure 1b is performed.

An output signal of the noise shaping device 205 is finally supplied to a digital/analog converter 206 which converts the digital output signal of the second noise shaping device 205 into an analog value.

The output signal of the digital/analog converter 206 is supplied to a post-filtering device 207 which filters out high frequency components which are caused by a stairstep-shaped variation of the analog output voltage

due to the digital/analog conversion processes in the digital/analog converter 206. A filtered output signal of the post-filtering device 207 is provided as an analog output signal 208 for further processing.

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The oversampling factor of 6.6, provided in this illustrative embodiment, and an interpolation filter unit 203 constructed as a second-order comb filter, a second first-order noise shaping device 205 and a  
10 digital/analog converter 206 which is constructed as a 10-bit current drive converter, result in an increase of a resolution by 2 bits, the circuit arrangement described with reference to figures 1a, b providing a saving in chip area by approximately a factor of 2.

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Although the present invention has been described above by means of preferred illustrated embodiments, it is not restricted to these but can be modified in many ways.

List of reference designations

In the figures, identical reference symbols designate identical or functionally identical components.

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100	Analog input signal
101	Analog/digital converter
102	Input terminal
103	Noise source terminal
104	Summing device
105	Digital signal
106	Output terminal
107	Decimation filter unit
110	Digital transmission signal
111	Quantization device
112	First noise shaping device
113	Noise signal
201	Mixing unit
202	Post-quantizing device
203	Interpolation filter unit
204	Amplifier unit
205	Second noise shaping device
206	Digital/analog converter
207	Post-filtering device
208	Analog output signal
209	Receive noise source terminal
210	Amplifier output signal
211	Receive noise signal